Quantized Output-feedback Control of Nonlinear Systems: A Cyclic-small-gain Approach

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Abstract: This paper presents a fundamentally novel approach to quantized output-feedback control of nonlinear systems. The advances made toward this challenging, yet important, problem are a seamless integration of dynamic quantization and the recently developed cyclic-small-gain theorem. The constructed output-feedback quantized controller guarantees the input-to-state stability (ISS) property for the closed-loop quantized system. More interestingly, an ISS-Lyapunov function is derived and is used in the implementation of output dynamic quantization, even with a 1-bit quantizer.

Key Words: Quantized feedback control, output-feedback, nonlinear systems, input-to-state stability (ISS), small-gain

1 INTRODUCTION

Over the past few years, quantized feedback control has gained intensive attention due to the convergence of controls and communications. Dynamic quantization was developed to deal with problems caused by the finite word length of the quantizers. The basic idea of dynamic quantization is to appropriately scale the quantization levels such that the input of the quantizer is always covered by the range of the quantizer, and at the same time, the state of the control system converges to the origin. Recent results of dynamically quantized control of networked systems and nonlinear systems can be found in [8, 10].

The input-to-state stability (ISS) methodology (see [15]) and ISS small-gain theorem [4, 3] have been introduced to quantized control to characterize the robustness of quantized control systems with respect to quantization errors in [10]. Quantized output feedback control has also been studied in [9, 11] for linear and nonlinear systems. Specifically, the authors of [11] presented an ISS-based framework for quantized output-feedback control.

This paper attempts to develop a fundamentally new design tool for quantized output-feedback control of nonlinear systems, which is general enough to recover previous studies in the recent literature. The novelty of the proposed approach lies in the employments of: (i) set-valued maps to cover the uncertainty caused by the quantization error; (ii) modified gain assignment technique [4] as a tool for control design; (iii) our recent result of cyclic-small-gain theorem [6, 12] to guarantee stability and to construct an ISS-Lyapunov function. Dynamic quantization is designed based on the ISS-Lyapunov function. Our result shows that the output of the quantized control system can be steered to within an arbitrarily small neighborhood of the origin even with a 1-bit uniform quantizer.

To make the paper self-contained, recall that a function \( \gamma : \mathbb{R}_+ \to \mathbb{R}_+ \) is positive definite if it is continuous, \( \gamma(s) > 0 \) for all \( s > 0 \) and \( \gamma(0) = 0 \). \( \gamma : \mathbb{R}_+ \to \mathbb{R}_+ \) is a \( \mathcal{K} \) function (denoted by \( \gamma \in \mathcal{K} \)) if it is continuous, strictly increasing and \( \gamma(0) = 0 \); it is a \( \mathcal{K}_\infty \) function (denoted by \( \gamma \in \mathcal{K}_\infty \)) if it is a \( \mathcal{K} \) function and also satisfies \( \gamma(s) \to \infty \) as \( s \to \infty \). \( \mathbf{I}_n \) represents the identity function. \( \mathbf{I}_n \) represents the identity matrix of size \( n \). \( \lambda_{\text{max}}(P) \) and \( \lambda_{\text{min}}(P) \) mean the largest and the smallest eigenvalues of a real and symmetric matrix \( P \), respectively.

2 PROBLEM FORMULATION

Consider the disturbed output-feedback nonlinear system with quantized output:

\[
\begin{align*}
\dot{x}_i &= x_{i+1} + f_i(y, d), & 1 \leq i \leq n - 1 \\
\dot{x}_n &= u + f_n(y, d) \\
y &= x_1 \\
y^q &= q(y, \mu)
\end{align*}
\]

where \( [x_1, \ldots, x_n]^T \in \mathbb{R}^n \) is the state, \( u \in \mathbb{R} \) is the control input, \( d \in \mathbb{R}^{m_d} \) represents external disturbance inputs, \( y \in \mathbb{R} \) is the output, \( q(y, \mu) \) is the output quantizer with variable \( \mu > 0 \). \( y^q \in \mathbb{R} \) is the quantized output, \( [x_2, \ldots, x_n]^T \) is the unmeasured portion of the state, and \( f_i \)’s (\( 1 \leq i \leq n \)) are uncertain locally Lipschitz continuous functions.

Remark 1 It is worth noting that the generality and importance of output-feedback nonlinear systems (1)–(3) has been well outlined in [7] and numerous references therein. However, little research has been done from a viewpoint of quantized feedback control due to technical challenges.

Output quantizer \( q(y, \mu) \) is defined as \( q(y, \mu) = \mu q_0 \left( \frac{y}{\mu} \right) \), where \( q_0 : \mathbb{R} \to \Omega \) is a piece-wise constant function satisfying \( |q_0(r) - r| \leq 1 \) if \( |r| \leq M \). We directly use the following property of \( q(y, \mu) \):

\[
|y| \leq M \mu \Rightarrow |q(y, \mu) - y| \leq \mu,
\]

where \( M > 0 \), \( M \mu \) is the quantization range and \( \mu > 0 \) is the maximum quantization error when \( |y| \leq M \mu \).
Remark 2 In [9], quantizers are formulated as $|y| \leq M \mu \Rightarrow |q(y, \mu) - y| \leq \Delta \mu$ with $\Delta > 0$. It is of interest to note that (5) is in accordance with the formulation in [9]. Indeed, if $\Delta \neq 1$, define $\mu' = \Delta \mu$, $M' = M/\Delta$ and $q'(y, \mu') = \mu'/(\Delta \mu')$. Then, $|y| \leq M' \mu' \Rightarrow |q'(y, \mu') - y| \leq \mu'$. Given fixed $M$, the basic idea of dynamic quantization is to dynamically update $\mu$ (and thus $M \mu$) in the process of quantized control to improve the control performance. Variable $\mu$ is referred to as zooming variable. Increasing $\mu$ enlarges the range and is called zooming-out; decreasing $\mu$ reduces the range and is called zooming-in.

In this paper, it is assumed that Assumptions 1–4 are made throughout the paper.

Assumption 1 System (1)–(3) with $u = 0$ is forward complete and small-time norm-observable with $y$ as the output.


Assumption 2 For each $f_i(y, d)$ ($1 \leq i \leq n$) in (1)–(2), there exists a known $\psi_{f_i} \in \mathcal{K}_\infty$ such that
\[ |f_i(y, d)| \leq \psi_{f_i}(|y|, |d^T|), \quad \forall y \in \mathbb{R}, \, \forall d \in \mathbb{R}^d. \] (7)

Assumption 3 The external disturbance $d$ in (1)–(2) is allowed to take values in $d \mathcal{B}$ with constant $d \geq 0$ and $\mathcal{B}$ the unit ball in $\mathbb{R}^d$.

Assumption 4 Quantizer $q(y, \mu)$ satisfies (5) with $M > 1$.

Remark 4 Assumption 4 is not restrictive. $M > 1$ means that quantization range $M \mu$ is strictly larger than the maximal quantization error $\mu$ when $|y| \leq M \mu$ This is satisfied even by the 1-bit uniform quantizer [8] shown in Fig. 1.

Definition 1 ([15]) For a continuous-time nonlinear system $\dot{\xi} = f(\xi, \varrho_1, \ldots, \varrho_m)$, with state $\xi \in \mathbb{R}^{n_\xi}$, external inputs $\varrho_i \in \mathbb{R}^{n_{\varrho_i}}$ ($i = 1, \ldots, m$) and locally Lipschitz vector field $f$, a function $V_\zeta : \mathbb{R}^{n_\xi} \to \mathbb{R}_+$ is an ISS-Lyapunov function if it is positive definite, radially unbounded and differentiable almost everywhere, and there exist $\chi^{\varrho_1}_\zeta, \ldots, \chi^{\varrho_m}_\zeta \in \mathcal{K}_\infty$ and $\alpha_\zeta$ positive definite such that
\[ V_\zeta(\xi) \geq \max_{i=1, \ldots, m} \{ \chi^{\varrho_i}_\zeta(|\varrho_i|) \} \Rightarrow \nabla V_\zeta(\xi)^T f(\xi, \varrho_1, \ldots, \varrho_m) \leq -\alpha_\zeta(V_\zeta(\xi)) \] (8)
holds wherever $V_\zeta$ is differentiable. Function $\chi^{\varrho_i}_\zeta$ is referred to as the ISS gain from $\varrho_i$ to $V_\zeta(\xi)$.

The objective of this paper is to design a quantized output feedback controller with dynamic quantization to steer the output $y$ to within an arbitrarily small neighborhood of the origin.

3 NEW TOOL FOR QUANTIZED OUTPUT- FEEDBACK CONTROL DESIGN

By convenience, denote $w = y^d - y$ as the quantization error.

3.1 Reduced-order Observer Design

We construct the following reduced-order observer:
\[ \dot{\hat{\xi}}_i = \xi_{i+1} + L_{i+1} y - L_i (\xi_2 + L_2 y^d), \quad 2 \leq i \leq n - 1 \] (9)
\[ \xi_n = u - L_n (\xi_2 + L_2 y^d) \] (10)
where $\xi_i$ is an estimate for the unmeasured state $x_i - Ly$ for each $2 \leq i \leq n$. Define $e_0 = [x_2 - L_2 y^d - \xi_2, \ldots, x_n - L_n y - \xi_n]^T$ as the observation error. Then, from (1)–(4) and (9)–(10), we get the observation error system:
\[ \dot{e}_0 = A e_0 + \phi_0(y, d, w) \] (11)
where
\[ A = \begin{bmatrix} -L_2 & \vdots & -L_{n-2} \\ \vdots & \ddots & \vdots \\ -L_n & \cdots & 0 \end{bmatrix}, \quad (12) \]
\[ \phi_0(y, d, w) = \begin{bmatrix} f_1(y, d) \\ \vdots \\ f_{n-1}(y, d) \\ L_2 - L_3 \\ \vdots \\ L_{n-1} L_2 - L_n \\ L_n L_2 \end{bmatrix} w. \] (13)

Real constants $L_i$‘s in (12) are chosen so that $A$ is Hurwitz, and thus there exists a matrix $P = P^T > 0$ satisfying $PA + A^T P = -2L_{n-1}$. For $e_0$ defined in (13), using Assumption 2, we can find $\psi^{(y)}_0, \psi^{(d)}_0, \psi^{(w)}_0 \in \mathcal{K}_\infty$ such that $|\phi_0(y, d, w)|^2 \leq \psi^{(y)}_0(|y|) + \psi^{(d)}_0(|d|) + \psi^{(w)}_0(|w|)$ holds for all $y, d, w$.

Define $V_0(e_0) = e_0^T P e_0$. Define $\alpha_0(s) = \lambda_{\min}(P)s^2$ and $\beta_0(s) = \lambda_{\max}(P)s^2$ for $s \in \mathbb{R}_+$. Then, $\beta_0(|e_0|) \leq V_0(e_0) \leq \alpha_0(|e_0|)$ holds for all $e_0$. Direct computation yields:
\[ \nabla V_0(e_0) = -2e_0^T P e_0 + 2e_0^T P \phi_0(y, d, w) \]
\[ \leq -e_0^T e_0 + |P|^2 |\phi_0(y, d, w)|^2 \]
\[ \leq - \frac{1}{\lambda_{\max}(P)} V_0(e_0) + |P|^2 (\psi^{(y)}_0(|y|) + \psi^{(d)}_0(|d|) + \psi^{(w)}_0(|w|)). \]
Define \( \chi_d^0 = 4\lambda_{\max}(P)|P|^2\psi_0^d \), \( \chi_h^0 = 4\lambda_{\max}(P)|P|^2\psi_0^h \), and \( \chi_b^0 = 4\lambda_{\max}(P)|P|^2\psi_0^b \). Then, we have
\[
V_0(e_0) \geq \max\{\chi_d^0(|y|), \chi_h^0(|d|), \chi_b^0(|w|)\}
\Rightarrow \nabla V_0(e_0) e_0 \leq -\alpha_0(V_0(e_0)) \tag{14}
\]
where \( \alpha_0(s) = \frac{1}{4\lambda_{\max}(P)}s \) for \( s \in \mathbb{R}_+ \).

### 3.2 Modified Gain Assignment Lemma

This subsection presents a modified version of the gain assignment technique in [4, 2, 14] for recursive quantized control design. Consider the following first-order system:
\[
\dot{\eta} = \phi(\eta, \delta_1, \ldots, \delta_m) + \nu
\]
where \( \eta \in \mathbb{R} \) is the state, \( \nu \in \mathbb{R} \) is the control input, and \( \delta_1, \ldots, \delta_m \in \mathbb{R} \) represent the disturbance inputs, the nonlinear function \( \phi(\eta, \delta_1, \ldots, \delta_m) \) satisfies
\[
|\phi(\eta, \delta_1, \ldots, \delta_m)| \leq \psi_{\phi}(|\eta, \delta_1, \ldots, \delta_m|)
\]
with \( \psi_{\phi} \in \mathcal{K}_{\infty} \). Define \( V_{\psi}(\eta) = \alpha_v(|\eta|) \) with \( \alpha_v(s) = \frac{1}{2} s^2 \) for \( s \in \mathbb{R}_+ \). Recall that \( \psi_{\phi} \) represents the standard sign function.

**Lemma 1** Consider system (15). For any constant \( 0 < c < 1, \epsilon > 0, \eta \geq 0 \) and \( \chi_\eta^0, \ldots, \chi_\eta^m \in \mathcal{K}_{\infty} \), one can find a smooth, odd, strictly decreasing and radially unbounded \( \kappa \) such that if
\[
v \in \{\kappa(\eta + a_0 \delta_{m+1} + \text{sgn}(\eta) \delta_{m+2}) : |a| \leq 1\}
\]
with \( \delta_{m+1}, \delta_{m+2} \in \mathbb{R}_+ \), then it holds that
\[
V_{\psi}(\eta) \geq \max_{k = 1, \ldots, m} \left\{ \chi_\eta^k(\delta_k), \alpha_v \left( \delta_{m+1} \right), c \right\}
\Rightarrow \nabla V_{\psi}(\eta) \eta \leq -\mu \nabla V_{\psi}(\eta).
\]

The proof of Lemma 1 is in Subsection 4.1.

**Remark 5** The existence of \( \delta_{m+2} \geq 0 \) in (17) does not influence the ISS property of the \( \eta \)-system. Intuitively, the term \( \text{sgn}(\eta) \delta_{m+2} \) with \( \delta_{m+2} \geq 0 \) consolidates the negative feedback strength of \( \eta \). See the proof of Lemma 1 for details.

### 3.3 Recursive Control Design

Define \( e_1 = y \). Consider the \( [\bar{e}_1^T, e_1, \xi_2, \ldots, \xi_n]^T \)-system:
\[
\begin{align*}
\dot{e}_0 &= A e_0 + \phi_0(e_0, e_1, d, w) \tag{19} \\
\dot{e}_1 &= \xi_2 + \phi_1(e_0, e_1, d) \tag{20} \\
\dot{\xi}_i &= \bar{e}_1 + \phi(e_1, \xi_2, w), \quad 2 \leq i \leq n - 1 \tag{21} \\
\dot{\xi}_n &= u + \phi_0(e_1, \xi_2, w) \tag{22}
\end{align*}
\]
where
\[
\begin{align*}
\phi_0(e_0, e_1, d) &= L_2 y + (x_2 - L_2 e_1 - \xi_2) + f_1(y, d) \\
\phi_1(e_1, \xi_2, w) &= L_1 y - \bar{e}_1 - L_3 (\xi_2 + L_2 y), \quad 2 \leq i \leq n - 1 \\
\phi_0(e_1, \xi_2, w) &= -L_1 (\xi_2 + L_2 y).
\end{align*}
\]

We get (20) from the \( x_1 \)-subsystem (1) using the fact that \( (x_2 - L_2 e_1 - \xi_2) \) is the first element of vector \( e_0 \). We get (21) and (22) from (9) and (10) by using \( y^d = y + w = e_1 + w \).

We will construct a new \( [\bar{e}_1^T, e_1, \ldots, e_n]^T \)-system consisting of ISS subsystems obtained through a recursive design of the \( [\xi_1^T, e_1, \xi_2, \ldots, \xi_n]^T \)-system. The ISS-Lyapunov function \( V_0 \) for the \( e_0 \)-subsystem is defined in Subsection 3.1. For \( 1 \leq i \leq n \), each \( e_i \)-subsystem will be designed with an ISS-Lyapunov function candidate
\[
V_i(e_i) = \alpha_v(|e_i|) \tag{23}
\]
where \( \alpha_v(s) = \frac{1}{2} s^2 \) for \( s \in \mathbb{R}_+ \). In the following discussions, we simply use \( \dot{V}_i \) instead of \( V_i(e_i) \) for \( 0 \leq i \leq n \). Denote \( \bar{e}_i = [e_0^T, e_1, \ldots, e_i]^T \) and \( \bar{\xi}_i = [\xi_2, \ldots, \xi_i]^T \).

In this subsection, we suppose that \( \mu \) is constant and consider only the case of \( |e_1| = |y| \leq M \mu \). From (5), this means \( |w| = |y^d - y| \leq \mu \).

#### 3.3.1 The \( e_0 \)-subsystem

Define \( \chi_d^0 = \chi_0^0 \circ \alpha_v^{-1} \) and \( \chi_b^0 = \chi_b^0 \). Then, from (14), we have
\[
V_0 \geq \max\{\chi_d^0(V_1), \chi_h^0(|d|), \chi_b^0(\mu)\}
\Rightarrow \nabla V_0 \dot{e}_0 \leq -\alpha_0(V_0).
\]

#### 3.3.2 The \( e_1 \)-subsystem

The \( e_1 \)-subsystem can be rewritten as
\[
\dot{\xi}_1 = \xi_2 - e_2 + (\phi_1(e_0, e_1, d) + e_2)
\]
\[
\dot{\xi}_2 = e_1 + \phi_1(e_1, \xi_2, w) \tag{25}
\]
with the new state variable \( e_2 \) to be defined below. From Assumption 2 and the definition of \( \phi_1 \), we can find a \( \psi^* \in \mathcal{K}_{\infty} \) such that \( (\phi_1(e_2, d), e_2^{*}) \leq \psi^*(|\bar{e}_i^T, d^T|^T) \).

Define a set-valued map \( S_1 \) as
\[
S_1(e_1, \mu) = \{k_1(e_1 + a_1 \mu) : |a| \leq 1\}
\]
with \( k_1 \) smooth, odd, strictly decreasing and radially unbounded, to be determined later. State variable \( e_2 \) is defined as
\[
e_2 = \begin{cases}
\xi_2 - \max S_1(e_1, \mu), & \text{if } \xi_2 \geq \max S_1(e_1, \mu) \\
\xi_2 - \min S_1(e_1, \mu), & \text{if } \xi_2 < \min S_1(e_1, \mu) \\
0, & \text{otherwise}
\end{cases}
\]
(27)

Then, we have \( \xi_2 - e_2 \in S_1(e_1, \mu) \).

For any \( \chi_1^0, \gamma_1^0 \in \mathcal{K}_{\infty} \), choose \( \chi_1^0 = \gamma_1^0 \circ \alpha_v^{-1} \) and \( \chi_1^0 = \gamma_1^0 \circ \alpha_v^{-1} \).

Then, \( \chi_1^0(V_0) = \chi_1^0 \circ \alpha_v^{-1}(V_0) \geq \chi_1^0(|e_0|) \) and \( \gamma_1^0(V_2) = \chi_1^0 \circ \alpha_v^{-1}(V_2) = \chi_1^0(|e_2|) \).

With Lemma 1, for any specified \( 0 < c_1 < 1, e_1 > 0, t_0 > 0, \gamma_1^0, \chi_1^0 \in \mathcal{K}_{\infty} \), we can find a smooth, odd, strictly decreasing and radially unbounded \( k_1 \) such that the \( e_1 \)-subsystem with \( \xi_2 - e_2 \in S_1(e_1, \mu) \) is ISS with \( V_1 \) satisfying
\[
V_1 \geq \max\{\chi_1^0(V_0), \gamma_1^0(V_2), \chi_1^0(|d|), \chi_1^0(\mu), e_1\}
\Rightarrow \nabla V_1 \dot{e}_1 \leq -t_1 V_1
\]
(28)
where \( \chi_1^0(s) = \alpha_v(s/c_1) \) for \( s \in \mathbb{R}_+ \).
**Remark 6** Since $\kappa_1$ is strictly decreasing, we can explicitly represent

\[
\begin{align*}
\max S_1(e_1, \mu) &= \kappa_1(e_1 - \mu), \\
\min S_1(e_1, \mu) &= \kappa_1(e_1 + \mu).
\end{align*}
\]

A set-valued map $S_1$ with smooth, odd, strictly decreasing and radially unbounded $\kappa_1$ and the definition of $e_2$ are shown in Fig. 2.

![Fig. 2: Bounds of set-valued map $S_1$ and the definition of $e_2$.](image)

**Remark 7** In standard backstepping [7], $\xi_2$ is usually considered as the virtual control of the $e_1$-subsystem. To take the quantization error $w$ into account, the virtual control law should be a function of $w$. However, the discontinuity of $w$ leads to discontinuity of the virtual control law, which is hard to handle by directly taking the derivative. The set-valued map $S_1$ defined in (26) overcomes this problem. As shown in Fig. 2, we can just steer $\xi_2$ to the region with smooth bounds $\max S_1$ and $\min S_1$, instead of driving $\xi_2$ to some specific discontinuous virtual control law.

### 3.3.3 The $e_i$-subsystem ($2 \leq i \leq n$)

When $3 \leq i \leq n$, for each $2 \leq k \leq i - 1$, suppose that

**H1** a set-valued map $S_k$ is defined as

\[
S_k(e_k, \xi_k, \mu) = \{\kappa_k(\xi_k - p_k) : p_k \in S_{k-1}(e_k, \xi_{k-1}, \mu)\}
\]

where $\kappa_k$ is smooth, odd, strictly decreasing and radially unbounded;

**H2** the new state variable $e_{k+1}$ is defined as

\[
e_{k+1} = \begin{cases} 
\xi_{k+1} - \max S_k(e_1, \xi_k, \mu), & \text{if } \xi_{k+1} \geq \max S_k(e_1, \xi_k, \mu); \\
\xi_{k+1} - \min S_k(e_1, \xi_k, \mu), & \text{if } \xi_{k+1} < \min S_k(e_1, \xi_k, \mu); \\
0, & \text{otherwise.}
\end{cases}
\]

**Remark 8** It is worth noting that, since $\kappa_k$ is strictly decreasing, it holds:

\[
\begin{align*}
\max S_k(e_1, \xi_k, \mu) &= \kappa_k(\xi_k - \max S_k(e_1, \xi_k - 1, \mu)), \\
\min S_k(e_1, \xi_k, \mu) &= \kappa_k(\xi_k - \min S_k(e_1, \xi_k - 1, \mu)).
\end{align*}
\]

**Lemma 2** Consider the $[\xi_1^T, e_1, \xi_2, \ldots, \xi_n]^T$-system in (19–22) with $|e_1| \leq M\mu$. For $2 \leq i \leq n$, with $S_1(e_1, \mu)$ and $e_2$ defined in (26) and (27) and Assumptions H1 and H2 satisfied for each $2 \leq k \leq i - 1$ if $3 \leq i \leq n$, for any variable $e_{i+1}$, when $e_i \neq 0$, we can derive the $e_i$-subsystem to

\[
e_i = \xi_{i+1} - e_{i+1} + \phi_i^* (e_{i+1}, d, \mu, w, \xi_i) \tag{33}
\]

where

\[
|\phi_i^* (e_{i+1}, d, \mu, w, \xi_i)| \leq \psi_i^* (e_{i+1}, d, \mu, w, \xi_i) \tag{34}
\]

with $\psi_i^* \in \mathcal{K}_\infty$. Specifically, $\xi_{i+1} = u$.

The proof of Lemma 2 is in Subsection 4.2. Define a set-valued map $S_i$ as

\[
S_i(e_1, \xi_i, \mu) = \{\kappa_i(\xi_i - p_i) : p_i \in S_{i-1}(e_1, \xi_{i-1}, \mu)\} \tag{35}
\]

with $\kappa_i$ smooth, odd, strictly decreasing and radially unbounded, to be defined later. Define $e_{i+1}$ as

\[
e_{i+1} = \begin{cases} 
\xi_{i+1} - \max S_i(e_1, \xi_i, \mu), & \text{if } \xi_{i+1} \geq \max S_i(e_1, \xi_i, \mu); \\
\xi_{i+1} - \min S_i(e_1, \xi_i, \mu), & \text{if } \xi_{i+1} < \min S_i(e_1, \xi_i, \mu); \\
0, & \text{otherwise.}
\end{cases}
\]

Then, we have $\xi_{i+1} - e_{i+1} \in S_i(e_1, \xi_i, \mu)$.

From the definition of $e_1$ (i.e., $e_{k+1}$ with $k = i - 1$) in Assumption H2, in the case of $e_i \neq 0$, for all $p_i \in S_{i-1}(e_1, \xi_{i-1}, \mu)$, it holds that $|\xi_i - p_i| \geq |e_i|$ and $\operatorname{sgn}(\xi_i - p_i) = \operatorname{sgn}(e_i)$, which means $\operatorname{sgn}(\xi_i - p_i - e_i) = \operatorname{sgn}(e_i)$, and thus $\xi_i - p_i = e_i + (\xi_i - p_i - e_i) = e_i + \operatorname{sgn}(e_i)(\xi_i - p_i - e_i)$.

Note that $\xi_{i+1} - e_{i+1} \in S_i(e_1, \xi_i, \mu)$. There always exists a $p_i \in S_{i-1}(e_1, \xi_{i-1}, \mu)$ such that $\xi_{i+1} - e_{i+1} = \kappa_i(\xi_i - p_i)$.

With Lemma 1, for any $e_i > 0$, $t_i > 0$, $\gamma^i_1, \gamma^i_2, \gamma^i_3, \gamma^i_4, \gamma^i_5 \in \mathcal{K}_\infty$, we can find a smooth, odd, strictly decreasing and radially unbounded $\kappa_i$ such that the $e_i$-subsystem with $\xi_{i+1} - e_{i+1} \in S_i(e_1, \xi_i, \mu)$ is ISS with $V_i$ satisfying

\[
V_i \geq \max_{k=0, \ldots, j-1,i+1} \{\gamma^i_k(V_i), \gamma^i_{k+1}(|d|), \gamma^i_{k+1}(|e_i|)\} \Rightarrow \forall V_i \dot{V}_i \leq -t_i V_i. \tag{37}
\]

By default, $V_{n+1} := \alpha_i(|e_{n+1}|)$. The true control input $u = \xi_{n+1}$ occurs with the $e_n$-subsystem, and we set $e_{n+1} = 0$.

### 3.3.4 Realizable Quantized Controller

From (37) with $i = n$, our desired quantized controller $u$ can be chosen in the following form:

\[
\begin{align*}
p_2^0 &= \kappa_1(y^0) \\
p_i^0 &= \kappa_{i+1}((\xi_{i-1} - p_{i-1}^0), \quad 3 \leq i \leq n \\
\end{align*}
\]

\[
u = \kappa_n(\xi_n - p_n^0). \tag{40}
\]

In the case of $|y| \leq M\mu$, we have $|w| = |y^0 - y| \leq \mu$ and thus $\kappa_1(y^0) = \kappa_1(y + w) = \kappa_1(e_1 + w) \in S_1(e_1, \mu)$. It is then directly checked that

\[
p_2^0 \in S_1(e_1, \mu) \Rightarrow \cdots \Rightarrow p_i^0 \in S_{i-1}(e_1, \xi_{i-1}, \mu) \Rightarrow \cdots \Rightarrow u = \xi_{n+1} - e_{n+1} \in S_n(e_1, \xi_n, \mu)
\]
where \( \epsilon_{n+1} = 0 \). Thus, if \( |y| \leq M_\mu \), then the quantized control law \((38)\)-(40) guarantees \((28)\) and \((37)\).

### 3.4 Cyclic-small-gain based Synthesis

Denote \( e = \bar{e}_n \). For \( 0 \leq i \leq n \), each \( e_i \)-subsystem has been made ISS (or more precisely, practically ISS). In this subsection, we fine tune the ISS-gains such that the \( e \)-system satisfies the cyclic-small-gain condition \([6, 12]\). The system graph of the \( e \)-system is shown in Fig. 3.

![Fig. 3: The system graph of the \( e \)-system.](image)

According to the recursive design, given the \( e_{i-1} \)-subsystem, by designing the set-valued map \( S_i \) for the \( e_i \)-subsystem, we assign the ISS gains \( \gamma_i \) \((1 \leq k \leq i - 1)\) such that

\[
\gamma_i^{k+1} \circ \gamma_i^{k+2} \circ \cdots \circ \gamma_i^{i-1} \circ \gamma_i^0 \circ \gamma_i^1 < \text{Id}.
\]

(41)

Applying this reasoning repeatedly, the \( e \)-system satisfies the cyclic-small-gain condition \([6, 12]\).

An ISS-Lyapunov function is constructed as:

\[
V(e) = \max_{0 \leq i \leq n} \{ \sigma_i(V(e_i)) \}
\]

(42)

with \( \sigma_i(s) = s \), \( \sigma_0(s) = \gamma_0 \circ \cdots \circ \gamma_{i-1}(s) \) \((2 \leq i \leq n)\) and \( \sigma_0(s) = \max_{0 \leq i \leq n} \{ \sigma_i \circ \rho_i^0(s) \} \) for \( s \in \mathbb{R}_+ \), where the \( \gamma_i^{(j)}\)'s are \( \mathcal{K}_\infty \) functions smooth on \((0, \infty)\) and slightly larger than the corresponding \( \gamma_i^{(j)}\)'s and still satisfy the cyclic-small-gain condition.

Recall \( |d| \leq d \). Denote \( \bar{e}_0 = 0 \). We represent the maximal influence of \( d \), \( \mu \), and \( e_i \) \((0 \leq i \leq n)\) as

\[
\vartheta = \max_{0 \leq i \leq n} \left\{ \sigma_i \circ \gamma_i^0(d), \sigma_i \circ \gamma_i^0(\mu), \sigma_i(e_i) \right\}.
\]

(43)

Using the Lyapunov-based cyclic-small-gain theorem in \([12]\), we achieve that if \( |y| \leq M_\mu \), then the \( e \)-system with quantized control law \((38)-(40)\) satisfies

\[
V(e) \geq \vartheta \Rightarrow \nabla V(e) \dot{e} \leq -\alpha(V(e))
\]

(44)

wherever \( \nabla V(e) \) exists, with \( \alpha \) positive definite. Note that \( \nabla V(e) \) exists almost everywhere \([12]\).

In the recursive design approach, we can make the \( \gamma_i^{(j)}\)'s \((0 \leq i \leq n, i \neq 1)\) arbitrarily small to get arbitrarily small \( \sigma_i\)'s \((0 \leq i \leq n, i \neq 1)\). We can also select the \( \chi_i^0\)'s \((0 \leq i \leq n, i \neq 1)\) arbitrarily small. In this way, for arbitrarily small \( \vartheta_0 > 0 \), we can design the gains such that \( \max_{0 \leq i \leq n} \{ \sigma_i \circ \gamma_i^0(d), \sigma_i(e_i) \} \leq \vartheta_0 \) and \( \max_{0 \leq i \leq n, i \neq 1} \{ \sigma_i \circ \chi_i^0(\mu) \} \leq \vartheta_0 \).

Recall \( \gamma_i^0(s) = \sigma_i(s/c_i) \) for \( s \in \mathbb{R}_+ \), defined in \((28)\). If \( |y| \leq M_\mu \), then quantized control law \((38)-(40)\) guarantees

\[
V(e) \geq \max \{ \alpha_q(\mu/c_1), \vartheta_0 \} \Rightarrow \nabla V(e) \dot{e} \leq -\alpha(V(e))
\]

(45)

wherever \( \nabla V(e) \) exists.

### 3.5 Dynamic Quantization

Define \( \Theta = \alpha_q(M_\mu) \). Then, \( \mu = \alpha_q^{-1}(\Theta)/M_\mu \). Dynamic quantization can be determined by designing an update law for \( \Theta \). Denote \( x = [x_1, \ldots, x_n]^T \) and \( \xi = [\xi_2, \ldots, \xi_n]^T \). Recall \( e = [e_0, \ldots, e_n]^T \) and the definition of \( e_i \) for \( i = 0, \ldots, n \).

The transformed state variable \( e \) can be considered as a continuous function of \( x, \xi, \mu \). Note that \( \mu = \alpha_q^{-1}(\Theta)/M_\mu \). Denote \( e = e(x, \xi, \Theta) \). In dynamic quantization, \( \Theta \) is piece-wise constant on the time-line and denoted as \( \Theta(t) \). Clearly, the piece-wise constant adjustment of \( \Theta \) leads to jumps of \( e \).

Due to space limitation, some results in this subsection are presented without proofs. However, the proofs are available from the authors upon request.

#### 3.5.1 Zooming-out Stage

The design of the zooming-out stage is motivated by \([11]\). In this stage, the control input \( u \) and the state \( \xi \) of the observer are set to be zero. The small-time norm-observability assumed in Assumption 1 guarantees that for \( d_i > 0 \), there exists a \( \varphi \in \mathcal{K}_\infty \) such that

\[
|\xi(t_k + d_i)| \leq \varphi(||\xi||_{t_k + d_i})
\]

(46)

for all \( k \in \mathbb{Z}_+ \). Considering the definitions of \( V \) and \( e \), for \( d_i > 0 \), there exists a \( \varphi \in \mathcal{K}_\infty \) such that

\[
|V(e(x(t_k + d_i), 0, 0))| \leq \varphi(||\xi||_{t_k + d_i})
\]

(47)

for all \( k \in \mathbb{Z}_+ \).

The forward completeness assumed in Assumption 1 guarantees that we can increase \( \Theta \) fast enough to dominate the growth rate of \( \varphi(||\xi||) \). Thus, we can design the zooming-out logic to increase \( \Theta \) (and thus \( \mu \)) fast enough such that at some time \( t_k > 0 \) with \( k^* \in \mathbb{Z}_+ \), it holds that

\[
\Theta(t_{k^*}) \geq \varphi(||\xi||_{t_{k^*-d_{t_k}}}) \geq \max \{ |V(e(x(t_{k^*}), 0, 0))| \}
\]

(48)

From the definition of \( S_i \) in \((26)\) and \((35)\), it can be observed that increase of \( \mu \) (and thus \( \Theta \)) leads to increase of max \( S_i \) and decrease of min \( S_i \). Using the definition of \( e_{i-1} \), increase of \( \Theta \) leads to decrease or hold of \( |e_{i-1}| \) (and thus decrease or hold of \( V(e) \)). Note that \( \xi(t_{k^*}) = 0 \). From \((48)\), we achieve

\[
\Theta(t_{k^*}) \geq \max \{ V(e(x(t_{k^*}), \xi(t_{k^*}), \Theta(t_{k^*})), 0) \}, \vartheta_0 \}
\]

(49)

#### 3.5.2 Zooming-in Stage

With the help of Assumption 4, in the constructive control design procedure, we can choose \( c_1 \) satisfying \( 1/M < c_1 < 1 \). Then, from the definition \( \Theta = \alpha_q(M_\mu) \), one can find a positive definite \( \rho_i^0 \) such that

\[
\alpha_q(\mu/c_1) \leq (\text{Id} - \rho_i^0(\Theta)).
\]

(50)

Suppose that at some time \( t_k > 0 \) with \( k \in \mathbb{Z}_+ \), it holds that

\[
\Theta(t_k) \geq \max \{ V(e(x(t_k), \xi(t_k), \Theta(t_k))), 0 \}
\]

(51)

We want to find a \( Q_{in}^0 : \mathbb{R}_+ \rightarrow \mathbb{R}_+ \) such that \( \Theta(t_{k+1}) = Q_{in}^0(\Theta(t_k)) \) satisfies

\[
\Theta(t_{k+1}) \geq \max \{ V(e(x(t_{k+1}), \xi(t_{k+1}), \Theta(t_{k+1}))), 0 \}
\]

(52)
where \( t_{k+1} - t_k = d_t \).

One can find a positive definite \( \rho_2^2 \) such that \( (\text{Id} - \rho_2^2) \in \mathcal{X}_0 \) and \( (\text{Id} - \rho_2^2)(s) \geq \max \{ (\text{Id} - \rho_1^2)(s), s - d_t \} \). Define

\[
\Xi = \text{Id} - \rho_2^2.
\]  

(53)

Condition (51) implies \( V(e(x(t_k), \xi(t_k), \Theta(t_k))) \leq \alpha_V(M \mu(t_k)) \). From (45) and (50), if (51) holds, then

\[
V(e(x(t_{k+1}), \xi(t_{k+1}), \Theta(t_{k+1}))) \leq \max \{ \Xi(\Theta(t_k)), \theta_0 \}.
\]  

(54)

Using the property of continuous functions, we can find a positive definite \( \rho_3^2 < \rho_2^2 \) such that for all \( x \in \mathbb{R}^n, \xi \in \mathbb{R}^{n+1} \), \( \Theta > 0 \) and \( h \geq 0 \), it holds that

\[
|V(e(x, \xi, \Theta - \rho_3^2(h))) - V(e(x, \xi, \Theta))| \leq h.
\]  

(55)

Define

\[
\Theta_m^0(\Theta) = \Theta - \rho_3^2 \left( \Theta - \max \left\{ \Xi(\Theta), \theta_0 \right\} \right).
\]  

(56)

Then, (54), (55) and (56) imply

\[
V(e(x(t_{k+1}), \xi(t_{k+1}), \Theta(t_{k+1}))) \leq \Theta(t_k) + \max \{ \Xi(\Theta(t_k)), \theta_0 \},
\]  

and (51) and (56) imply

\[
\Theta(t_{k+1}) \geq \frac{\Theta(t_k) + \max \{ \Xi(\Theta(t_k)), \theta_0 \}}{2} \geq \theta_0.
\]  

(57)

(58)

Properties (57) and (58) together guarantee (52).

Lemma 3 Suppose that \( \Theta(t_k) \geq \theta_0 \) with \( k \in \mathbb{Z}_+ \). Then, the zooming-in logic \( \Theta(t_{k+1}) = \Theta_m^0(\Theta(t_k)) \) for \( k \in \mathbb{Z}_+ \) with \( \Theta_m^0 \) defined in (56) guarantees

\[
\lim_{{k \to \infty}} \Theta(t_k) = \theta_0.
\]  

(59)

The proof of Lemma 3 is in Subsection 4.3. The motions of \( \Theta(t) \) and \( V(e(x(t), \xi(t), \Theta(t))) \) in the zooming-in stage are shown in Fig. 4.

![Fig. 4: Motions of \( \Theta(t) \) and \( W(t) = V(e(x(t), \xi(t), \Theta(t))) \) in the zooming-in stage.](image)

With the appropriately designed zooming-in logic in (56), it always holds that \( V(e(x(t), \xi(t), \Theta(t))) \leq \Theta(t) \) in the zooming-in stage. Thus, the closed-loop signals are bounded. By using (59), we have \( \lim_{{t \to \infty}} V(e(x(t), \xi(t), \Theta(t))) \leq \theta_0 \). Recall the definition of \( V \) in (42). It can be observed that \( \gamma = x_1 = e_1 \) ultimately converges to within the region \( |y| \leq \psi_0^{-1}(\theta_0) \). By choosing \( \theta_0 \) arbitrarily small, output \( y \) can be steered to within an arbitrarily small neighborhood of the origin. Recall \( \Theta = \alpha_V(M \mu) \). With \( \Theta_m^0 \) defined in (56), the zooming-in logic for \( \mu \) is designed as

\[
Q_m(\mu) = \frac{1}{M} \alpha_V^{-1} \circ \Theta_m^0 \circ \alpha_V(M \mu).
\]  

(60)

3.6 Main Result

The main result of this paper is summarized in Theorem 1.

Theorem 1 Consider system (1)–(4) with output quantization satisfying (5). Under Assumptions 1–4, the closed-loop signals are bounded, and in particular, the output \( y \) can be steered to within an arbitrarily small neighborhood of the origin with the quantized output-feedback controller composed of reduced-order observer (9)–(10), control law (38)–(40), and dynamic quantization of form (6) with zooming-in dynamics \( Q = Q_m \) defined in (60).

4 PROOFS OF LEMMAS 1–3

4.1 Proof of Lemma 1

Using (16), one can find \( \psi_\phi^0, \psi_\phi^1, \ldots, \psi_\phi^m \in \mathcal{X}_\infty \) such that

\[
|\phi(\eta, \delta_1, \ldots, \delta_m)| \leq \psi_\phi^0(|\eta|) + \sum_{k=1}^{m} \psi_\phi^k(|\delta_k|).
\]  

(61)

Note that \( \psi_\phi^0(s) + \sum_{k=1}^{m} \psi_\phi^k(\chi_\eta^{-1}) \circ \alpha_V(s) + \frac{1}{2} s \) is a \( \mathcal{X}_\infty \) function of \( s \). With Lemma 1 in [2], for any \( 0 < c < 1 \) and \( \varepsilon > 0 \), one can find a \( \rho : \mathbb{R} \to \mathbb{R}^+ \), positive, nondecreasing and smooth on \( (0, \infty) \), such that

\[
(1 - c) \rho((1 - c)^2 s^2) s \geq \psi_\phi^0(s) + \sum_{k=1}^{m} \psi_\phi^k(\chi_\eta^{-1}) \circ \alpha_V(s) + \frac{1}{2} s \geq 0
\]  

(62)

for all \( s \geq \sqrt{2} \). Define \( \kappa(\tau) = -\rho(\tau^2) r \) for \( r \in \mathbb{R} \). Then, \( \kappa \) is smooth, odd, strictly decreasing and radially unbounded. Recall \( V_\eta(\eta) = \alpha_V(|\eta|) = \frac{1}{2} |\eta|^2 \). Consider the case of \( V_\eta(\eta) \geq \max_{\delta_1, \ldots, \delta_m} \left\{ \chi_\eta^{-1}(\delta_k), 0 \right\}, \alpha_V(\delta_k) = \frac{\delta_k - \eta}{c \eta} \). We have

\[
|\delta_k| \leq \chi_\eta^{-1} \circ \alpha_V(|\eta|), \quad 1 \leq k \leq m
\]  

(63)

\[
\delta_{m+1} \leq c \alpha_V^{-1}(V_\eta(\eta)) = c |\eta|
\]  

(64)

\[
|\eta| \geq \sqrt{2} \epsilon
\]  

(65)

The \( v \) satisfying (17) can be represented as \( v = \kappa(\eta^2) = -\rho(\eta^2) \eta^2 \) where \( \eta^2 = \eta + \text{sgn}(\eta) \delta_{m+2} + a \delta_{m+1} \) with \( |a| \leq 1 \). With \( 0 < c < 1 \) and (64), when \( \eta \neq 0 \), we have \( \text{sgn}(\eta) = \text{sgn}(\eta), |\eta^2| \geq |\eta| + a \delta_{m+1} \geq (1 - c)|\eta| \) and thus \( \rho(\eta^2)|\eta| \geq (1 - c) \rho((1 - c)^2 \eta^2)|\eta| \). From (61)–(65) and the discussion above, we obtain

\[
\begin{align*}
\nabla V_\eta(\eta)(\phi(\eta, \delta_1, \ldots, \delta_m) + v) &= \nabla \eta(\phi(\eta, \delta_1, \ldots, \delta_m) - \rho(\eta)^2 \eta) \\
&\leq |\eta| \nabla \eta(\phi(\eta, \delta_1, \ldots, \delta_m) - |\eta| \rho(\eta^2) \eta^2) \\
&\leq |\eta| \nabla \eta^1(|\eta|) + \sum_{k=1}^{m} \psi_\phi^k(\chi_\eta^{-1}) \circ \alpha_V(|\eta|) \\
&\quad - (1 - c) \rho((1 - c)^2 \eta^2)|\eta| \\
&\leq - \frac{1}{2} |\eta|^2 = -V_\eta(\eta).
\end{align*}
\]  

(66)

This ends the proof.
4.2 Proof of Lemma 2

We simply use $S_k$ instead of $S_k(e_i, \xi_k, \mu)$ for $1 \leq k \leq i - 1$. We only consider the time instants at which $e_i > 0$. The proof for the time instants at which $e_i < 0$ is similar.

Consider the definition of $S_1$ in (26) and the iteration-type definitions of $S_k$'s in Assumption H1. Recall (29), (30), (31) and (32). The smoothness of the $K$-th implies the smoothness of max$S_1$ with respect to $e_1$ and the smoothness of max$S_k$ with respect to $\xi_k$ and max$S_{k-1}$ for $2 \leq k \leq i - 1$. Repeatedly using the property of the composition of smooth functions, we can see max$S_1$ is smooth with respect to $[e_1, \xi_{i-1}]^T$ and thus $\max S_1 - \min S_1$ is continuous with respect to $[e_1, \xi_{i-1}]^T$. In the case of $e_i > 0$, the dynamics of $e_i$ can be derived as

$$\dot{e}_i = \frac{\partial \max S_{i-1}}{\partial [e_1, \xi_{i-1}]^T} [e_1, \xi_{i-1}]^T = \xi_{i+1} + \phi(e_i, \xi_2, w) - \frac{\partial \max S_{i-1}}{\partial [e_1, \xi_{i-1}]^T} [e_1, \xi_{i-1}]^T$$

$$:= \xi_{i+1} - e_{i+1} + \phi^*(\dot{e}_i, 1, d, \mu, w, \bar{w})$$

(67)

Specifically, $\xi_{i+1} = u$. We used (20), (21) and (22) to get the last equality above.

Recall $[e_1] \leq M \mu \Rightarrow |w| \leq \mu$. From (20), (21) and (22), we can see $|\phi(e_1, \xi_2, w)|$ is bounded by a $C^\infty$ function of $|[e_1, \xi_2, \mu]^T|$, and $|[\dot{e}_1, \xi_{i-1}, \mu]^T|$ is bounded by a $C^\infty$ function of $|[\dot{e}_1, \xi_{i-1}, \mu]^T|$. Thus, we achieve that $|\phi^*(\dot{e}_i, 1, d, \mu, w, \bar{w})|$ is bounded by a $C^\infty$ function of $|[\dot{e}_i, \mu]^T|$. As pointed out in [13], this can be achieved by proving $|\dot{e}_i, \mu|^T = 0 \Rightarrow \xi = 0$. By using the definitions of $S_1$ and $e_2$ in (26) and (27) and the definitions of $S_k$ and $e_{k+1}$ in Assumptions H1 and H2, we have

$$S_1 = \{0\} \Rightarrow \xi_2 = 0 \Rightarrow \xi_3 = 0 \Rightarrow \cdots \Rightarrow \xi_k = 0$$

$$S_k = \{0\} \Rightarrow \xi_{k+1} = 0 \Rightarrow e_{k+1} = 0, \quad 2 \leq k \leq i - 1.$$

This ends the proof.

4.3 Proof of Lemma 3

Consider the following two cases.

(a) $\Xi(\Theta(t_k)) \geq \bar{\theta}_0$. From the definition of $\Xi$ in (53), one can find a positive definite function $\rho^*_s$ such that $\rho^*_s(1) \geq \rho^*_s(s)$ for $s \in \mathbb{R}^+$. In the case of $\Xi(\Theta(t_k)) \geq \bar{\theta}_0$, we have

$$\Theta(t_{k+1}) = \Theta(t_k) - \rho^*_s \left( \frac{\Theta(t_k) - \Xi(\Theta(t_k))}{2} \right) \leq \Theta(t_k) - \rho^*_s(\Theta(t_k))$$

(68)

which guarantees that there exists a $t_{w} > 0$ with $k \in \mathbb{Z}_+$ such that $\Xi(\Theta(t_{w})) < \bar{\theta}_0$, and equivalently, $\Theta(t_{w}) < \Xi^{-1}(\bar{\theta}_0)$.

(b) $\Xi(\Theta(t_k)) < \bar{\theta}_0$. Define $\Theta'(t_k) = \Theta(t_k) - \bar{\theta}_0$ for $k \in \mathbb{Z}_+$. Then, we obtain

$$\Theta'(t_{k+1}) = \Theta'(t_k) - \rho^*_s \left( \frac{\Theta'(t_k)}{2} \right)$$

(69)

which is a asymptotically stable first-order discrete-time system [5], and implies $\lim_{t \to \infty} \Theta(t) = 0$.

Recall the definition of $\Xi$ in (53). We can see $\Xi^{-1} > Id$ and $\Xi^{-1}(\bar{\theta}_0)$ is larger than $\bar{\theta}_0$. Considering both cases (a) and (b), we have $\lim_{t \to \infty} \Theta(t_k) = \bar{\theta}_0$. This ends the proof.

5 CONCLUSIONS

This paper makes significant progress on the challenging, yet important, problem of quantized output-feedback control of nonlinear systems, by developing a new tool based on recent cyclic-small-gain techniques [6, 12]. The result shows that dynamic quantization can be implemented even with a 1-bit uniform quantizer. Furthermore, the influence of the external disturbance can be attenuated to an arbitrarily small level. It is our firm belief that the proposed design tool will prove helpful for quantized feedback control of other important classes of nonlinear systems.

REFERENCES


